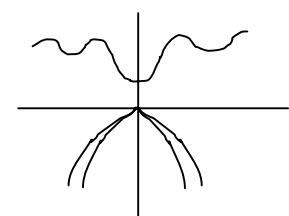


半導体の有効質量

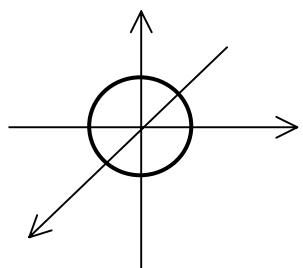


直接遷移型半導体の電子の有効質量

$\mathbf{k}_x, \mathbf{k}_y, \mathbf{k}_z$ 全ての方向に同じ有効質量

等エネルギー面は k 空間で Γ 点（原点）を中心には

$$\text{球状} \quad E - E_c = \frac{\hbar^2(k_x^2 + k_y^2 + k_z^2)}{2m_e^*}$$

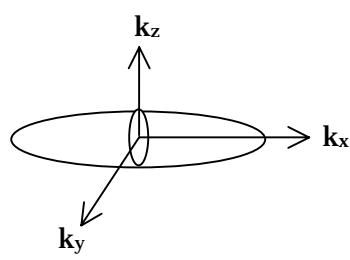


間接遷移型半導体の有効質量

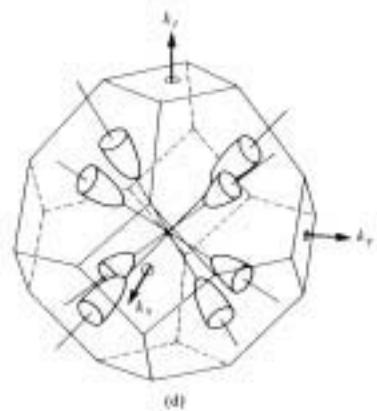
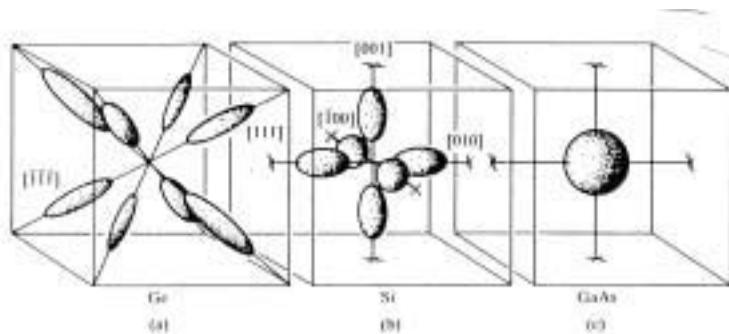
Si, Ge の場合

$$E - E_c = \frac{\hbar^2}{2m_\ell^*}k_x^2 + \frac{\hbar^2}{2m_t^*}(k_y^2 + k_z^2)$$

$$\frac{k_x^2}{\alpha^2} + \frac{k_y^2 + k_z^2}{\beta^2} = 1$$



$$\alpha = \sqrt{\frac{2m_\ell^*(E - E_c)}{\hbar^2}} \quad \beta = \sqrt{\frac{2m_t^*(E - E_c)}{\hbar^2}}$$



$$N_{el} \left(\frac{4}{3} \pi \alpha \beta^2 \right) = \frac{4}{3} \pi k_{eff}^3 \Rightarrow N_{el} (m_l^* m_t^{*2}) = (m_e^*)^{\frac{3}{2}}$$

ブリリュアンゾーン中の
葉巻の数 Si で 6、Ge で 4

$m_e^* = 6^{\frac{2}{3}} (m_l^* m_t^{*2})^{\frac{1}{3}}$ が Si
 $m_e^* = 4^{\frac{2}{3}} (m_l^* m_t^{*2})^{\frac{1}{3}}$ が Ge

電子の状態密度有効質量 m_{de}^* (状態の数)

電子の電気伝導度有効質量 m_{ce}^* は間接遷移では異方的



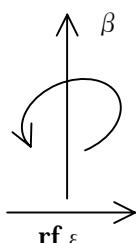
正孔の有効質量 m_h^*

$$g_V(E) = \frac{\sqrt{2} m_h^{*\frac{3}{2}} (E_V - E)^{\frac{1}{2}}}{\pi^3 \hbar^3} = \frac{\sqrt{2} m_{hh}^{*\frac{3}{2}} (E_V - E)^{\frac{1}{2}}}{\pi^3 \hbar^3} + \frac{\sqrt{2} m_{lh}^{*\frac{3}{2}} (E_V - E)^{\frac{1}{2}}}{\pi^3 \hbar^3}$$

よって $(m_h^*)^{\frac{3}{2}} = (m_{hh}^*)^{\frac{3}{2}} + (m_{lh}^*)^{\frac{3}{2}}$

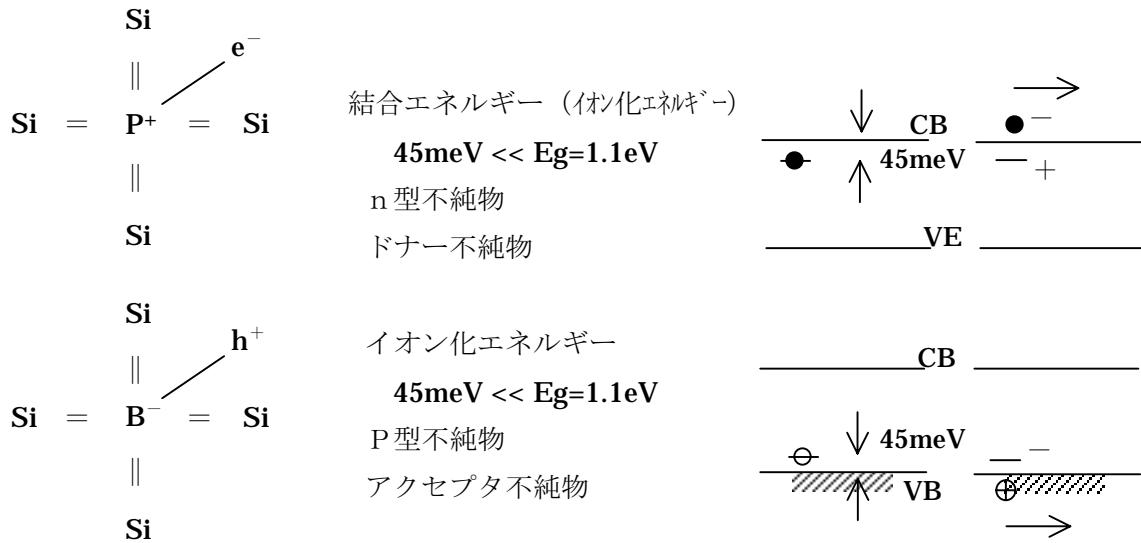
$$m_h^* = \left[(m_{hh}^*)^{\frac{3}{2}} + (m_{lh}^*)^{\frac{3}{2}} \right]^{\frac{2}{3}}$$

有効質量の測定法 サイクロトロン共鳴



有効質量	Ge	Si	GaAs
m_l^* / m_0	1.588	0.19163	—
m_t^* / m_0	0.08152	0.1905	—
m_e^* / m_0	—	—	0.067
m_{hh}^* / m_0	0.347	0.537	0.51
m_{lh}^* / m_0	0.0429	0.153	0.082
m_{so}^* / m_0	0.077	0.234	0.154

不純物半導体



GaAs 中の C は？

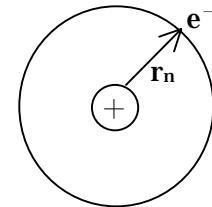
- ボーワ原子（浅い不純物）

$$\text{角運動量} = m^* v r_n = n \hbar \dots \dots \textcircled{1}$$

$$\text{SI 単位系で } \frac{m^* v^2}{r_n} = \frac{e^2}{4\pi\epsilon\epsilon_0 r_n^2} \dots \dots \textcircled{2}$$

$\Rightarrow 8.85 \times 10^{-12}$

$$\textcircled{1}-\textcircled{2} \text{ より } r_n = \frac{4\pi\epsilon\epsilon_0 (n\hbar)^2}{m^* e^2} \dots \dots \text{n=1} \quad \text{ボーワ半径} \quad \dots \dots \textcircled{3}$$



Ge	$r_1 \sim 8\text{nm}$
Si	$r_1 \sim 3\text{nm}$
GaAs	$r_1 \sim 12\text{nm}$

電子の運動エネルギー(E_k)とポテンシャルエネルギー(U)は

$$E_k = \frac{1}{2} m^* v^2 = \frac{1}{2} \left(\frac{e^2}{4\pi\epsilon\epsilon_0 r_n} \right)$$

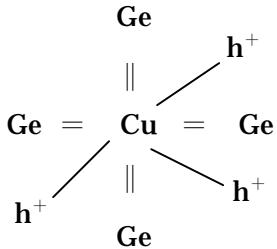
$$U = -\frac{e^2}{4\pi\epsilon\epsilon_0 r_n} \quad (r = \infty \text{ で } U = 0)$$

$$\text{よって、 } E_n = E_k + U = -\frac{1}{2} \left(\frac{e^2}{4\pi\epsilon\epsilon_0 r_n} \right) \dots \dots \textcircled{4}$$

$$\textcircled{3} \rightarrow \textcircled{4} \quad E_n = -\frac{m^* e^4}{2(4\pi\epsilon\epsilon_0 n\hbar)^2} = \frac{13.6}{\epsilon^2} \frac{m^*}{m_0} \quad (\text{eV})$$

大体の半導体で $\epsilon \sim 10$ $m^* \leq m_0 \Rightarrow E_n \leq 0.1\text{eV}$

深い準位



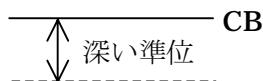
$$\left\{ \begin{array}{l} \text{Cu}^0 (\text{中性の Cu}) \text{は } 3 \text{ 個のホールを有する} \\ \text{Cu}^{-1} \quad \quad \quad \text{は } 2 \text{ 個} \\ \text{Cu}^{-2} \quad \quad \quad \text{は } 1 \text{ 個} \\ \text{Cu}^{-3} \quad \quad \quad \text{は } 0 \text{ 個} \end{array} \right.$$

$$\text{イオン化エネルギー: } E^{0/-} < E^{-1/-} < E^{-2/-} >> E_1(\text{Ge}) \approx 10 \text{ meV}$$

33meV

深い準位

実空間で局在している
 $\Delta \mathbf{x}$ が小さい

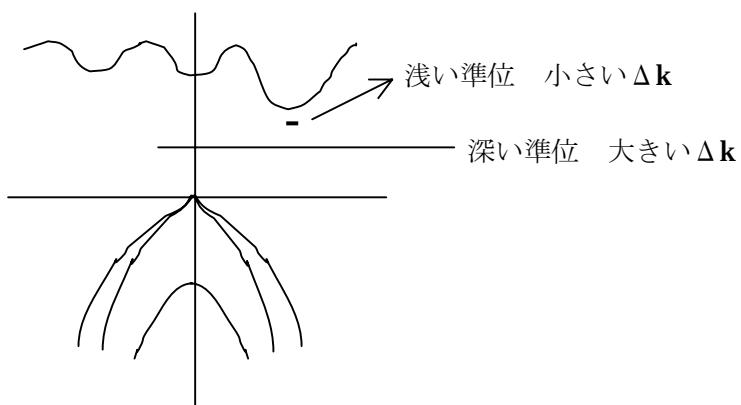


$E - k$ のバンド図では?

$$\Delta p \Delta x \sim h$$

$$\hbar \Delta k \Delta x \sim h$$

$$\Delta k \Delta x \sim 2\pi$$



$\text{Si} = \text{Si} = \text{Si}$

$\parallel \quad \parallel \quad \parallel$ Au は Si 内の深い準位

$\text{Si} = \text{Si} = \text{Si}$

$\parallel \quad \parallel \text{(Au)} \parallel$

$\text{Si} = \text{Si} = \text{Si}$

$$n_i^2 = np$$

中性の条件

$\rightarrow \text{pn}$ 接合



縮退

$$p - n + N_D^+ - N_A^- = 0 \rightarrow \text{トランジスター}$$

\rightarrow ホール効果

SiC の例

フェルミ レベルの位置

$$\text{真性半導体} \quad n = p \rightarrow N_C e^{(E_F - E_C)/kT} = N_V e^{(E_V - E_F)/kT}$$

ここで $E_F = E_i$ とおくと

$$E_i = \frac{E_C + E_V}{2} + \frac{kT}{2} \ln \left(\frac{N_V}{N_C} \right) \quad \Rightarrow \quad \frac{N_V}{N_C} = \left(\frac{m_h^*}{m_e^*} \right)^{\frac{3}{2}}$$

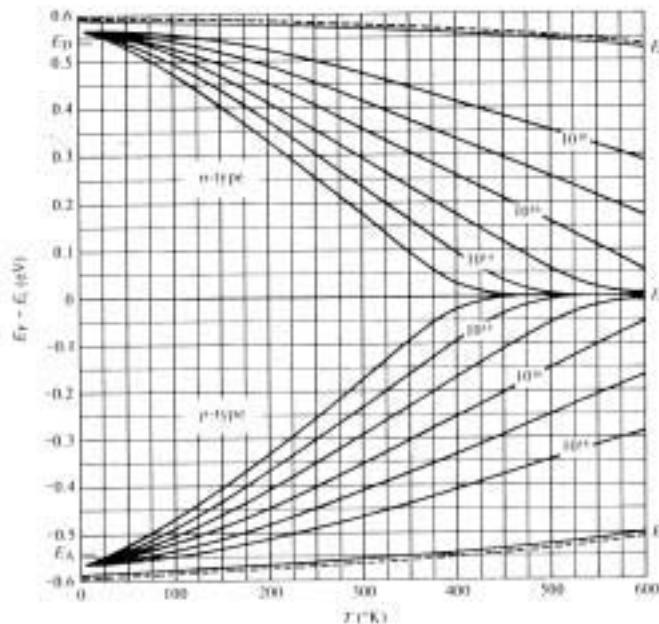
$$E_i = \frac{E_C + E_V}{2} + \frac{3}{4} kT \ln \left(\frac{m_h^*}{m_e^*} \right)$$

不純物半導体 $kT > E_D, E_A$ (ドナー、アクセプタのイオン化エネルギー)

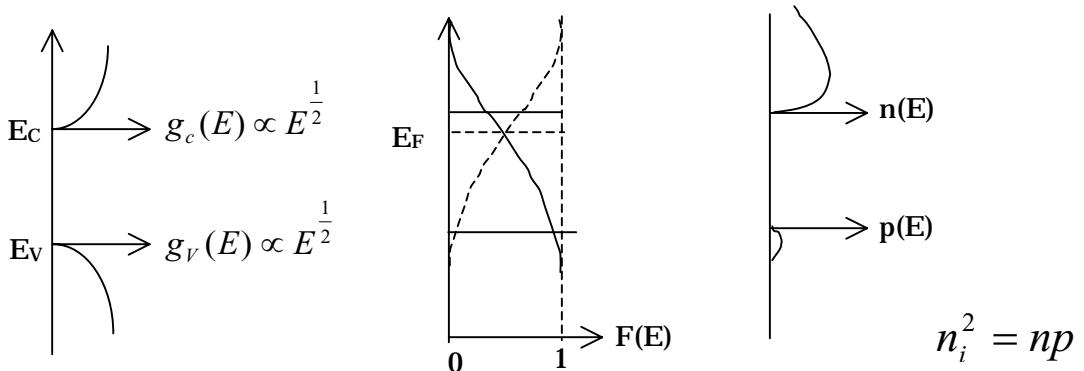
$n = N_D$ (ドナー濃度) 又は $p = N_A$ (アクセプタ濃度)

$$E_F - E_i = kT \ln \left(\frac{n}{n_i} \right) = -kT \ln \left(\frac{P}{n_i} \right)$$

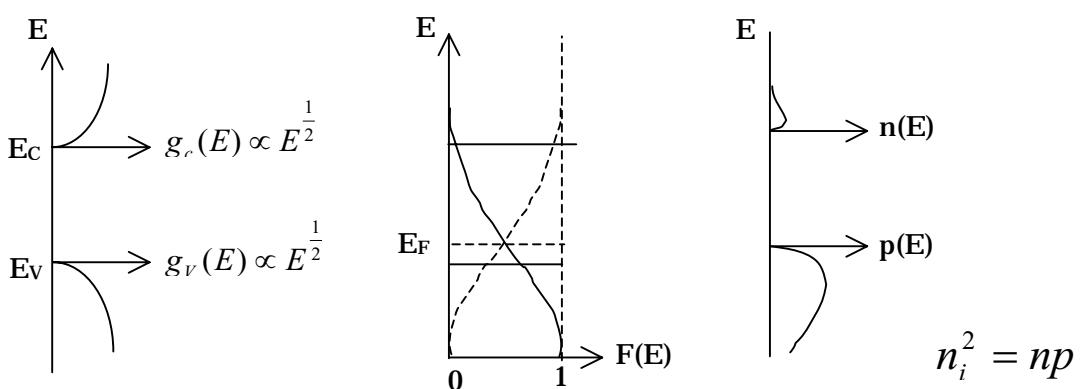
$$\begin{cases} E_F - E_i = kT \ln \left(\frac{N_D}{n_i} \right), \quad N_D \gg N_A, N_D \gg n_i \\ E_i - E_F = kT \ln \left(\frac{N_A}{n_i} \right), \quad N_A \gg N_D, N_A \gg n_i \end{cases}$$



n型

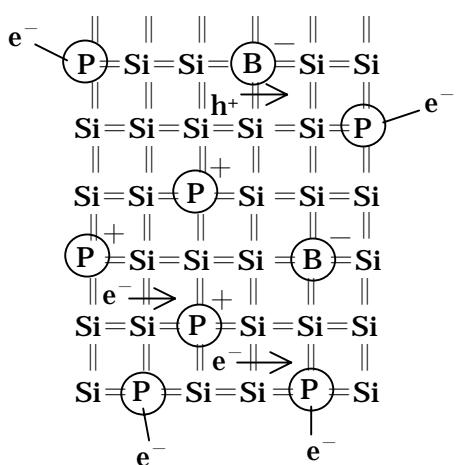


p型



電荷中性の法則

浅い準位の例で考える



$$N_D = 7$$

$$N_A = 2$$

左の例では $N_D^+ = 3$, $N_A^- = 2$

$$n=2, p=1$$

$$\begin{cases} \text{マイナスの数} & n + N_A^- = 4 \\ \text{プラスの数} & p + N_D^+ = 4 \end{cases}$$

等しくなるので中性

$$p - N + N_D^+ - N_A^- = 0$$

pn 接合

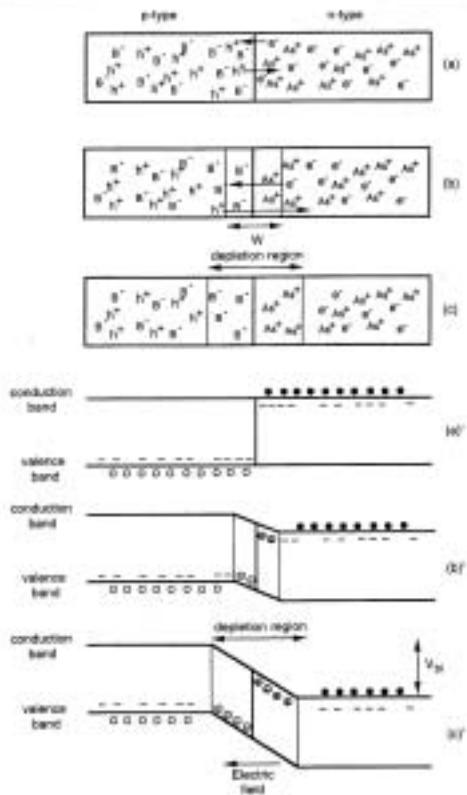


Fig. 2 pn-junctions and their band diagrams

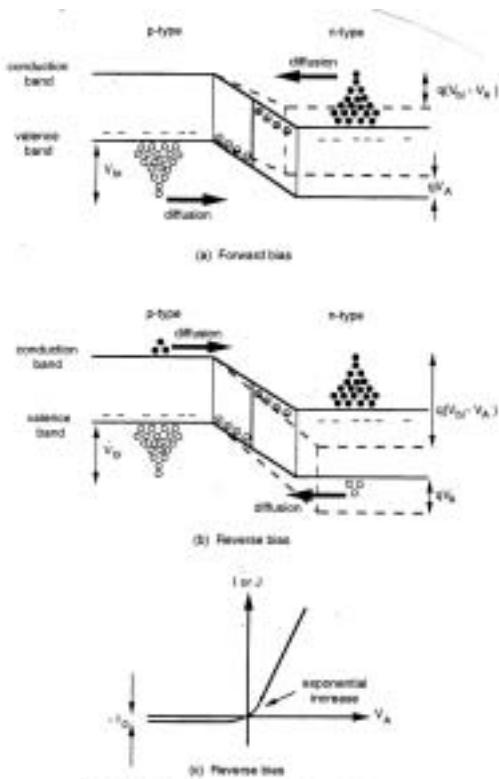


Fig. 3 Current flow mechanisms in pn-junctions

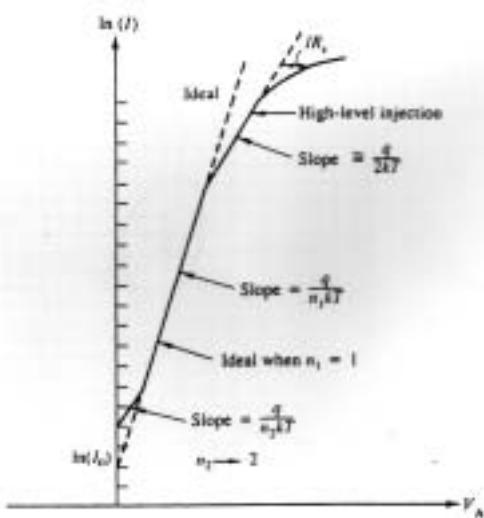


Fig. 4 順バイアスにおける電流・電圧特性（文献 1 より転写）

Bipolar transistor

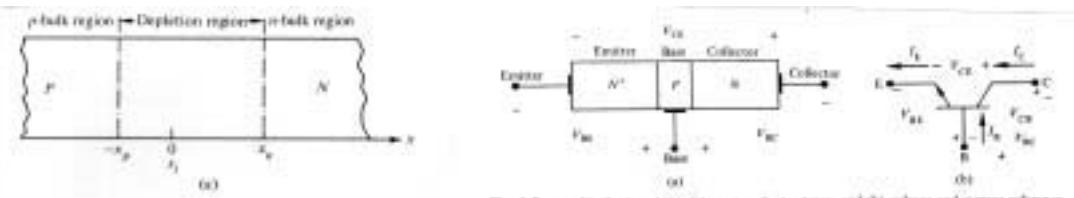


Fig. 1.2 a) Bipolar transistor: (a) semiconductor types; and (b) voltage and carrier reference polarities

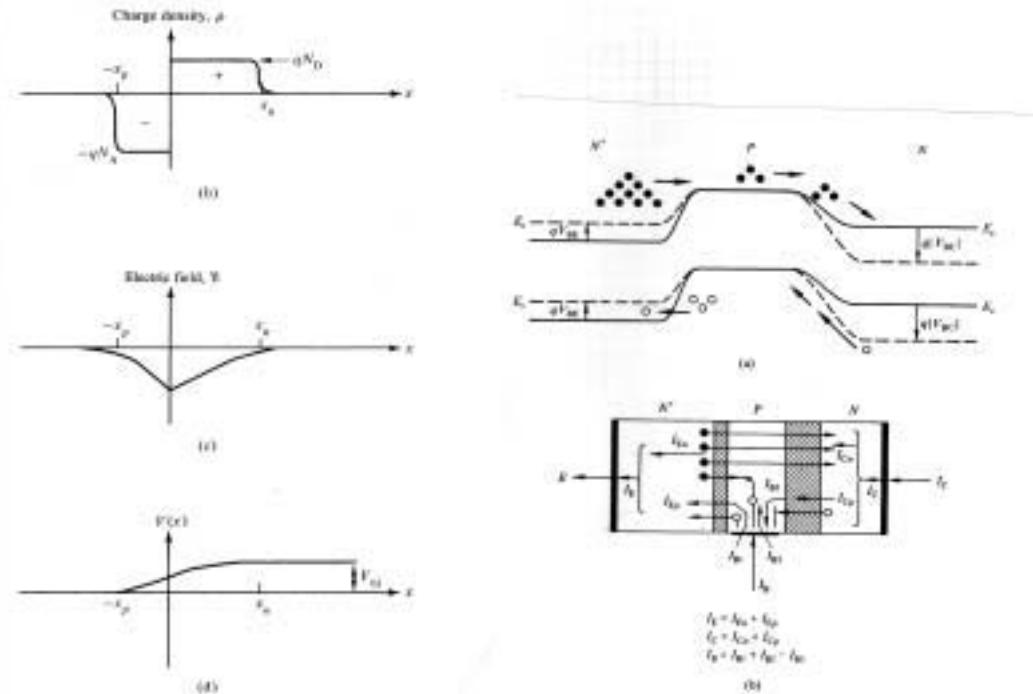


Fig. 2.3 Depletion region electrostatics.

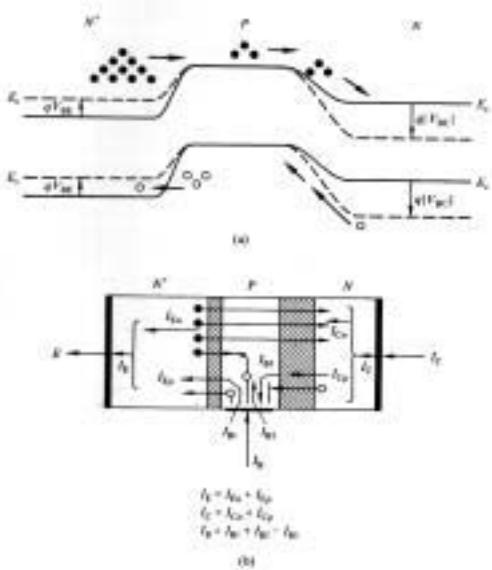


Fig. 1.7 n-p-n in the active region: (a) energy band diagram (— thermal equilibrium; - - active region); (b) current components and carrier flux.

JFET

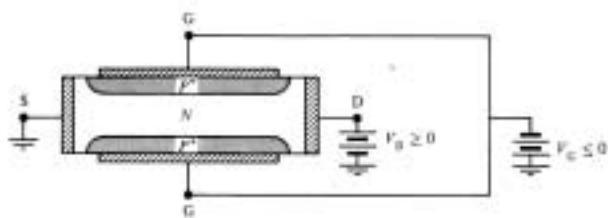


Fig. 1.4 Specification of the device structure and biasing conditions assumed in the qualitative analysis.

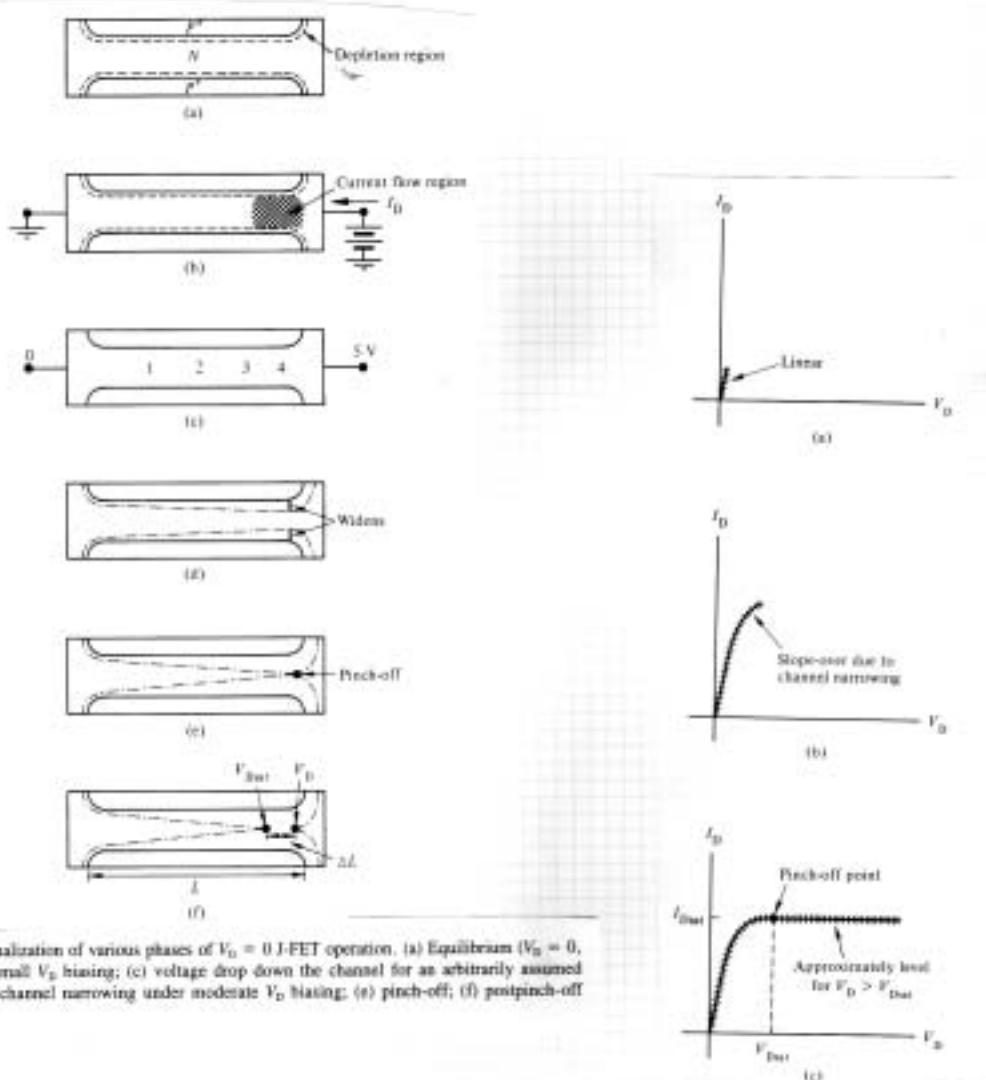


Fig. 1.5 Visualization of various phases of $V_D = 0$ J-FET operation. (a) Equilibrium ($V_D = 0$, $V_G = 0$); (b) small V_D biasing; (c) voltage drop down the channel for an arbitrarily assumed $V_D = 5V$; (d) channel narrowing under moderate V_D biasing; (e) pinch-off; (f) postpinch-off ($V_D > V_{Dsat}$).

Fig. 1.6 General form of the I_D - V_D characteristics. (a) Linear, simple resistor, variation for very small drain voltages. (b) Slope-over at moderate drain biases due to channel narrowing. (c) Pinch-off and saturation for drain voltages in excess of V_{Dsat} .

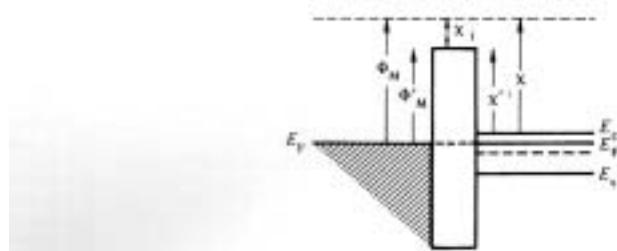


Fig. 2.3 Equilibrium energy band diagram for an ideal MOS structure.

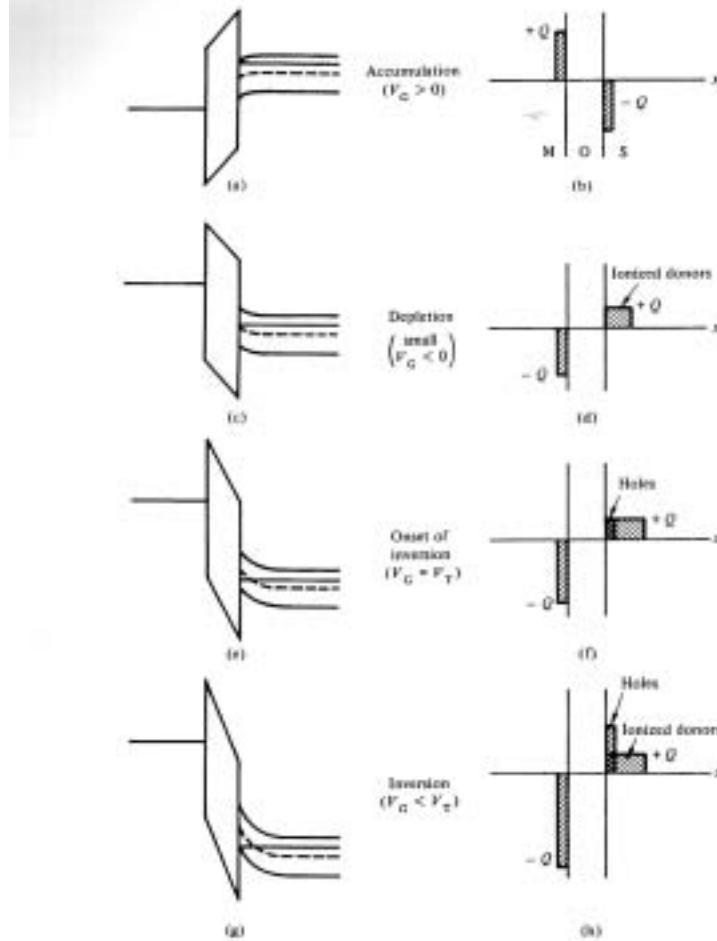


Fig. 2.5 Energy band diagrams and corresponding block charge diagrams describing the static state in an ideal n-type MOS-capacitor.

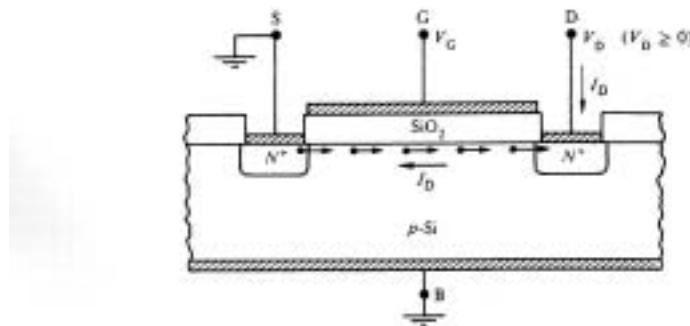


Fig. 3.1 Cross-sectional view of the basic p-bulk (n-channel) MOSFET structure showing the terminal designations, carrier and current flow directions, and standard biasing conditions.

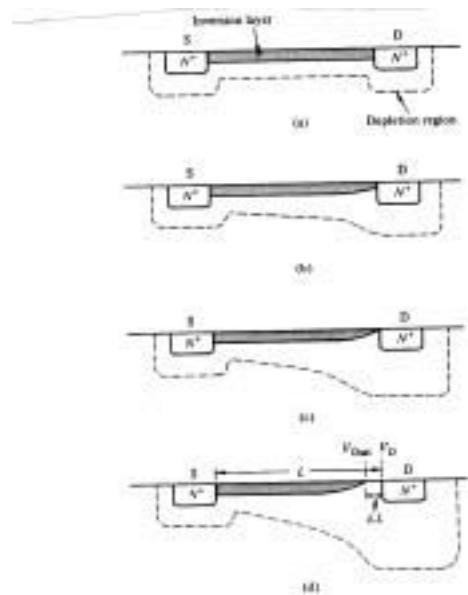


Fig. 3.2. Visualization of various phases of $V_G > V_T$ MOSFET operation. (a) $V_G = 0$; (b) channel (inversion layer) narrowing under moderate V_D biasing; (c) pinch-off; and (d) pinched-off ($V_D > V_{D_{on}}$) operation. (Note that the inversion layer widths, depletion widths, etc. are not drawn to scale.)

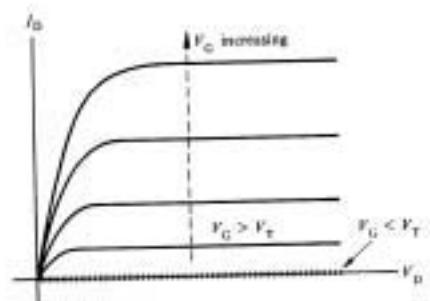
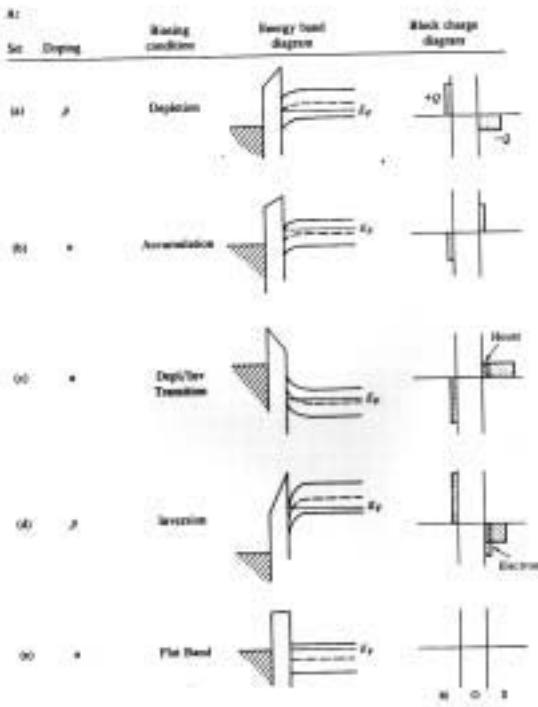
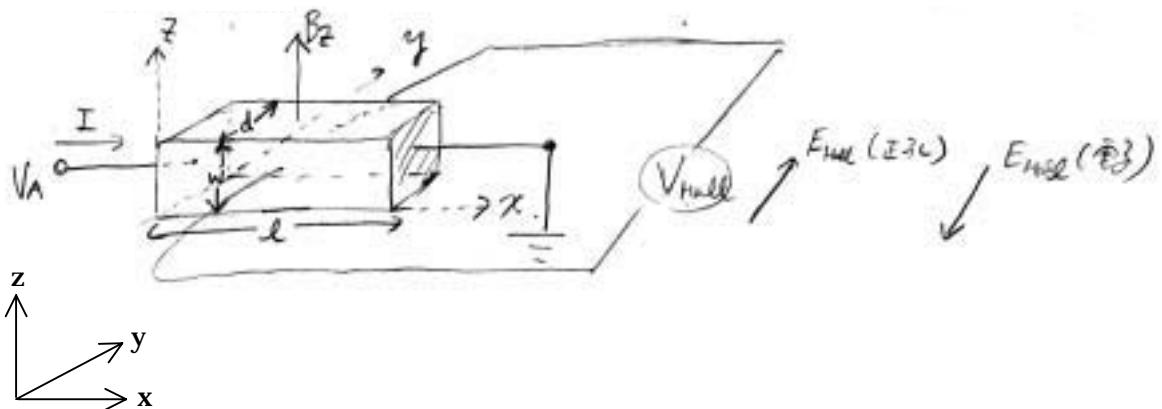


Fig. 3.4. General form of the I_D - V_D characteristics expected from a long channel ($\Delta L \ll L$) MOSFET.



ホール効果



$$\vec{F} = -e\vec{v} \times \vec{B} = -e\mu\vec{E} \times \vec{B}$$

$$F_y = -e\mu E_x B_z$$

平衡状態では $-eE_{Hall} = -F_y = e\mu E_x B_z$

電子の場合 $E_{Hall} = -\mu_x E_x B_z$

ホール係数

$$\begin{aligned} R_H &= \frac{E_{Hall}}{J_x B_z} = \frac{V_{Hall}/d}{(I/wd)B} = \frac{V_H W}{BI} \\ &= \frac{E_{Hall}}{\sigma_x E_x B_z} = \frac{-\mu B_z E_x}{\underbrace{\mu_x en E_x B_z}_{\sigma_x}} = -\frac{1}{en} \end{aligned}$$

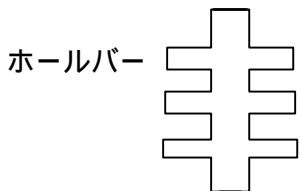
$$R_H = \frac{V_{Hall}W}{BI} \quad \left\{ \begin{array}{l} = -\frac{1}{en} (\text{電子}) \\ = \frac{1}{en} (\text{正孔、ホール}) \end{array} \right.$$

$$\text{さらに } \sigma = \begin{cases} en\mu \\ ep\mu \end{cases}$$

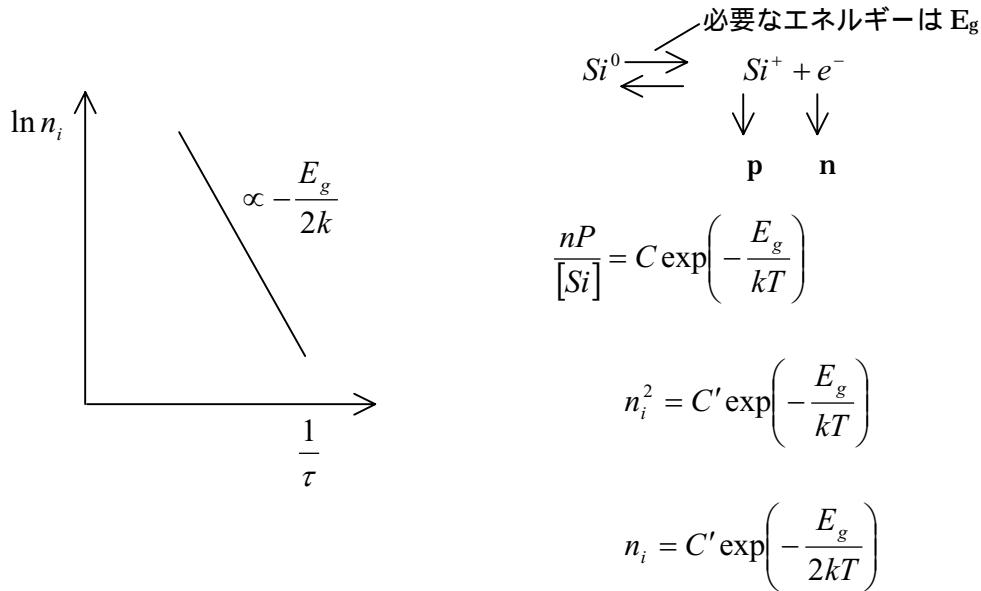
だから 抵抗測定から 求め

ホール効果から n

移動度 μ も計算できる。

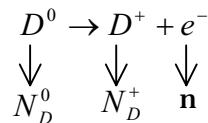


真性半導体のホール効果測定効果



では不純物半導体では？

ドナー濃度 $N_D >> 0$ で考える。 アクセプタ濃度 $N_A = 0$



$$\frac{N_D^+ n}{N_D^0} = C_D \exp\left(-\frac{E_D}{kT}\right)$$

$$\text{ここで } N_D^+ = n \quad N_D^0 = N_D - n$$

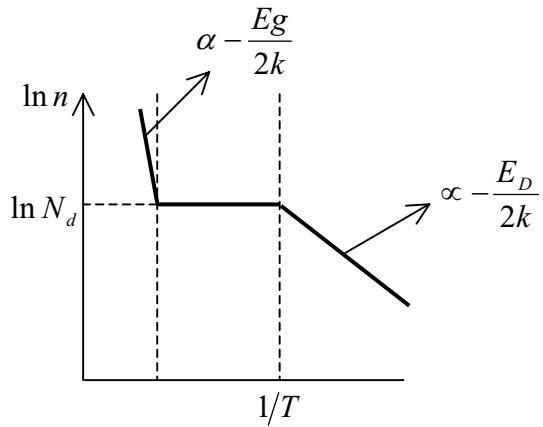
$$\frac{n^2}{N_D - n} = C_D \exp\left(-\frac{E_D}{kT}\right)$$

$$\text{高温領域 } kT \gg E_g \gg E_D \rightarrow n_i \gg N_D \quad n \propto \exp\left(-\frac{E_g}{2kT}\right)$$

$$\text{中温領域 } E_g \gg kT \gg E_D \rightarrow N_D \approx N_D^+ \quad n \approx N_D$$

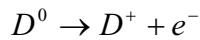
$$\text{低温領域 } E_D > kT \quad N_D \gg n \quad \text{よって}$$

$$\frac{n^2}{N_D - n} = C_D \exp\left(-\frac{E_D}{kT}\right) \Rightarrow n = C'_D \exp\left(-\frac{E_D}{2kT}\right)$$



補償のある半導体では？

N_D と N_A が共存している。ただし、 $N_D \gg N_A$ (n型) を仮定すると



は同じ

$$N_D^0 \quad N_D^+ \quad \mathbf{n}$$

$$\frac{N_D^+ n}{N_D^0} = C_{DA} \exp\left(-\frac{E_D}{kT}\right)$$

ここで $\begin{cases} N_D^+ = n + N_A \\ N_D^0 = N_D - N_D^+ = N_D - n - N_A \end{cases} \longrightarrow \begin{cases} \text{すべてのアクセプタが-にイオン化 } N_A^- = N_A \\ \text{同じ量のドナーが+にイオン化 } N_D^+ = N_A \end{cases}$

$$\frac{(n + N_A)n}{N_D - n - N_A} = C_{DA} \exp\left(-\frac{E_D}{kT}\right)$$

$$\text{高温領域} \quad kT \gg E_g \gg E_D \rightarrow n_i \gg N_D \quad n \propto \exp\left(-\frac{E_g}{kT}\right)$$

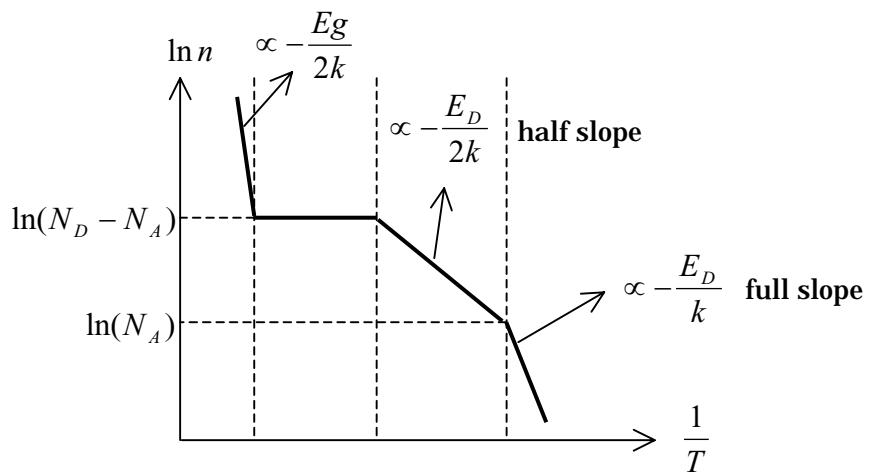
$$\text{中温領域} \quad E_g \gg kT \gg E_D \rightarrow n = N_D - N_D \quad \text{一定}$$

$$\text{低温領域} \quad E_D \gg kT \quad N_D \gg n \gg N_A$$

$$\frac{(n + N_A)n}{N_D - n - N_A} = \frac{n^2}{N_D} = C_{DA} \exp\left(-\frac{E_D}{kT}\right) \Rightarrow n = C'_{DA} - \exp\left(-\frac{E_D}{2kT}\right)$$

$$\text{超低温領域} \quad E_D \gg kT \quad N_D \gg N_A \gg n$$

$$\frac{(n + N_A)n}{N_D - n - N_A} = \frac{nN_A}{N_D} = C''_{DA} \exp\left(-\frac{E_D}{kT}\right) \Rightarrow n = C''_{DA} \exp\left(-\frac{E_D}{kT}\right)$$

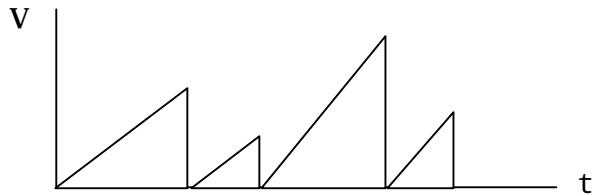


移動度の変化は？

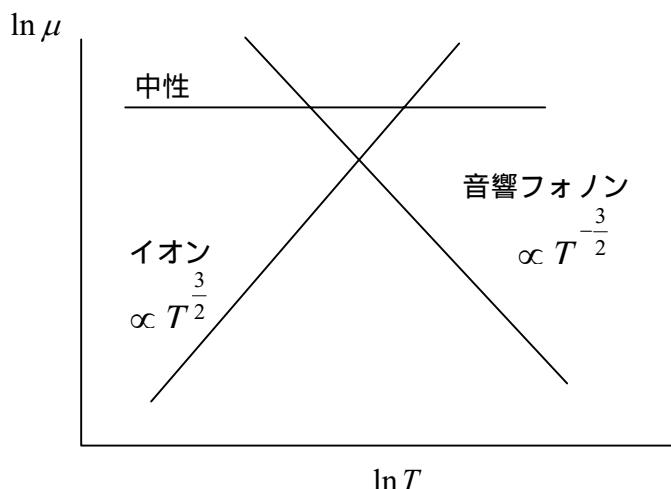
$$\mu = \frac{e\langle\tau\rangle}{m^*}$$

$\langle\tau\rangle$: 平均緩和時間

散乱の要因



- 音響フォノン散乱 $\langle\tau\rangle \propto T^{-\frac{3}{2}}$
- イオン化不純物散乱 $\langle\tau\rangle \propto T^{\frac{3}{2}}$
- 中性不純物散乱 $\langle\tau\rangle$ は T に依存しない
- 光学フォノン散乱・インターバレー散乱 $\langle\tau\rangle \propto T^{-2.3}$ 位



トータルな移動度 μ_{tot}

$$\frac{1}{\mu_{tot}} = \frac{1}{\mu_{\text{音響フォノン}}} + \frac{1}{\mu_{\text{イオン}}} + \frac{1}{\mu_{\text{中性}}}$$